



Comparative Study of Optimal Self Tuning Fuzzy PID, Fuzzy and PID Controller for Speed Control of BLDC Motor

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الخلاصة

في هذا البحث تمت دراسة ومقارنة أنواع مختلفة من تقنيات السيطرة للسيطرة على سرعة محرك التيار المستمر. المسيطرات التي تم تصميمها هي عبارة عن مسيطر تقليدي ونوعين من المسيطرات الذكية. المسيطر الأول الذي تم تصميمه هو المسيطر التناصي التكاملـي. المسيطر الثاني هو المسيطر الضبابي حيث يعطي نظام سيطرة مستقر إذ يتكون من ادخالين وإخراج واحد وان دوال العضوية لكل من الإدخال والإخراج هي سبع دوال من نوع المثلث. أخيراً، تم تصميم الضبابي الأمثل ذاتي الضبط لبارامترات المسيطر التناصي التكاملـي التفاضلي للسيطرة على سرعة محرك التيار المستمر حيث يقوم بتضييق قيم الكسب (KP, KI & KD). الإخراج لنظام السيطرة الضبابي يمثل بaramترات المسيطر التكاملـي التفاضلي. تمت مقارنة استجابات السرعة لمحرك التيار المستمر وللمسيطرات الثلاثة. نتائج المحاكاة تبين ان المسيطر الذي تم اقتراحه يعطي استجابة أفضل مقارنة بالمسيطرات الأخرى. كذلك فان للمسيطر المقترن القدرة للوصول إلى السرعة الهدف والمتانة في التغلب اضطرابات العزم المخارجي.

الكلمات المفتاحية

المسيطر المنطقي الضبابي لمحرك BLDC، المسيطر PID، STF PID، سيطرة السيرعة لـ BLDC.

can controlled the speed of DC motors manually or automatically by using many controllers type [4].

2. Problem Formulation

The target of this research is to tune of the PID parameters by using various intelligent methods for concentration control of dc motor speed and comparative the different for three controllers [5]. The BLDC motors are the best for wide scales speed and therefore used in many drives adjustable [6]. To make BLDC motor model similar to PID transfer function must converted it to a single transfer function as shown in equation.1[7]. Fig.(1). shows the BLDC motor mathematical model[8].

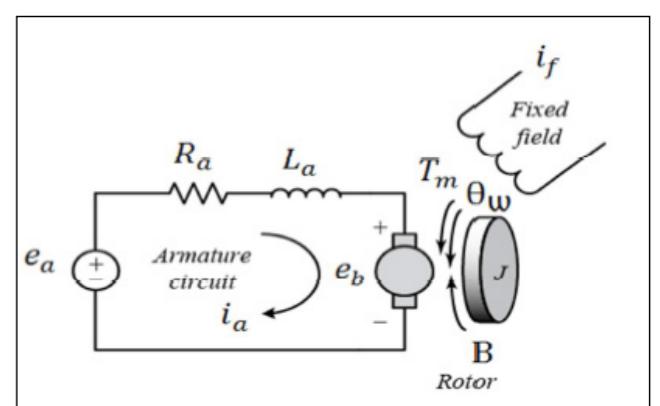


Fig. (1): Model of BLDC Motor

$$\frac{\omega(s)}{v(s)} = \frac{Kt}{Ljs^2 + (D + Rj)s + KtKb} \quad \text{--- (1)}$$

Table I shows the BLDC motor parameters and their values:

Table (I):Parameters of BLDC Motor

Parameter	Values and units	Parameters	Values and units
J: Rotor inertia	10^{-5} Kgm s ² /rad	D :viscous friction constant	10^{-4} Kg-m s/rad
) k _b : EMF constant	0.1434 Vs rad- m/a	L: electric resistance	0.054 H
R: electric resistance	12 Ω	Kt: Torque constant	0.1433 Kg-Nm/A

Fig. (2) shows the response of the DC motor speed in the open loop circuit system state and applied an external torque amount of 20Nm at time equal 0.5 sec. Where we note that the speed broke down when the external torque applied, there for we need to design of the control system to increase the stability of speed when applied external torque, as well as access the required reference speed.

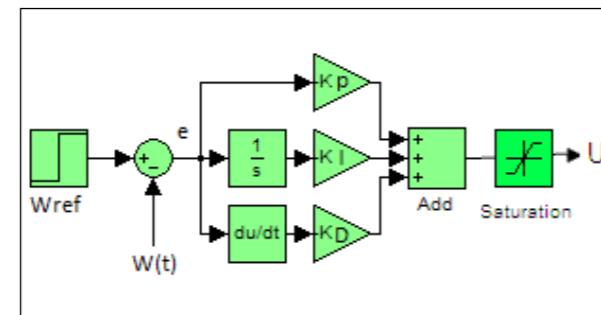


Fig. (3): Simulink Model PID Controller

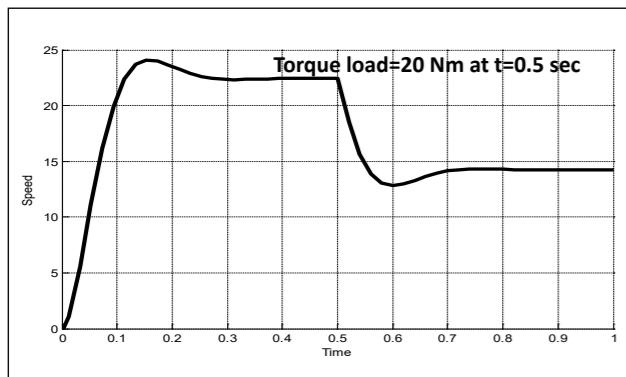


Fig. (2):Open loop speed response of the DC motor

4. Conventional Pid Design

The PID controller is a feedback type conventional controller which stands for Proportional-Integral-Derivative [9]. The error (e) between the target set point and the current speed is used to generate proportional, integral and derivative action[10]. We applied Ziegler-Nichols method to design the PID controller. Fig. (3) shows the Simulink model of the PID controller. The transfer function used for PID controller is [11]:

$$U = (\text{Proportional gain} + \text{Integral gain} + \text{Derivative gain}) \\ = (K_p) + (K_i/s) + (K_d s) \quad \dots \dots \dots (2)$$

5. Conventional Fuzzy Logic

The main structure parts of the FLC are a fuzzification part, a fuzzy logic reasoning part, a knowledge base, and a defuzzification part. In this paper we have two inputs fuzzy logic control, 7-membership function of each inputs (e and ch_e) & outputs. To design the fuzzy controller first should be created rule base in FIS Editor. It was subsequently tuning gain values for each input ($G(e)$, $G(ch-e)$) and output, $G(u)$,as well as adjust the membership functions and the amount of overlap between them. Fig. (4) shows the simulation model of fuzzy logic design.

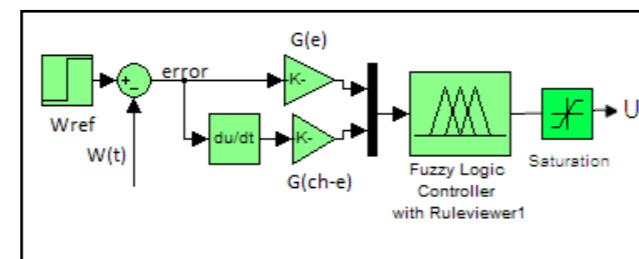


Fig. (4): Fuzzy Logic Controller

The triangular membership functions of these fuzzy sets for error & change of error are cleared with seven triangular membership functions. The membership functions is shown in Fig. (5). We notice that the , (L,H ,Z ,s, m,b) as low, high, near zero, small, medium and b big respectively. , Lm stands for low-medium, Hb for high big, and so on [7].

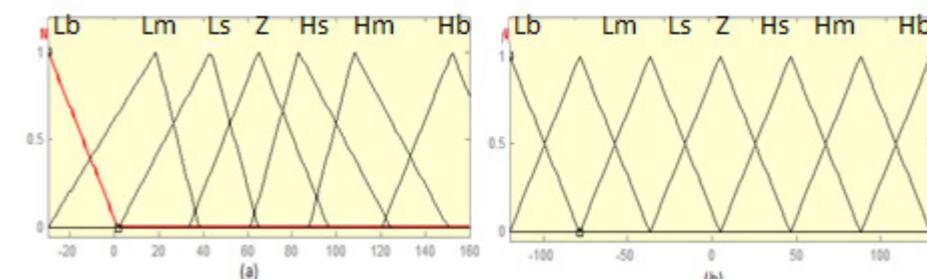


Fig. (5): FLC membership functions of input (a- error & b- ch-error)

Table (2): Fuzzy control

error/ch-error	<u>Lb</u>	<u>Lm</u>	<u>Ls</u>	<u>Z</u>	<u>Hs</u>	<u>Hm</u>	<u>Hb</u>
<u>Lb</u>	<u>Lb</u>	<u>Lb</u>	<u>Lb</u>	<u>Lb</u>	<u>Lm</u>	<u>Ls</u>	<u>Z</u>
<u>Lm</u>	<u>Lb</u>	<u>Lb</u>	<u>Lb</u>	<u>Lm</u>	<u>Ls</u>	<u>Z</u>	<u>Hs</u>
<u>Ls</u>	<u>Lb</u>	<u>Lb</u>	<u>Lm</u>	<u>Ls</u>	<u>Z</u>	<u>Hs</u>	<u>Hm</u>
<u>Z</u>	<u>Lb</u>	<u>Lm</u>	<u>Ls</u>	<u>Z</u>	<u>Hs</u>	<u>Hm</u>	<u>Hb</u>
<u>Hs</u>	<u>Lm</u>	<u>Ls</u>	<u>Z</u>	<u>Hs</u>	<u>Hm</u>	<u>Hb</u>	<u>Hb</u>
<u>Hm</u>	<u>Ls</u>	<u>Z</u>	<u>Hs</u>	<u>Hm</u>	<u>Hb</u>	<u>Hb</u>	<u>Hb</u>
<u>Hb</u>	<u>Z</u>	<u>Hs</u>	<u>Hm</u>	<u>Hb</u>	<u>Hb</u>	<u>Hb</u>	<u>Hb</u>

Then system is run and the 3-dimensional behavior of the FLC is shown in fig.(6).

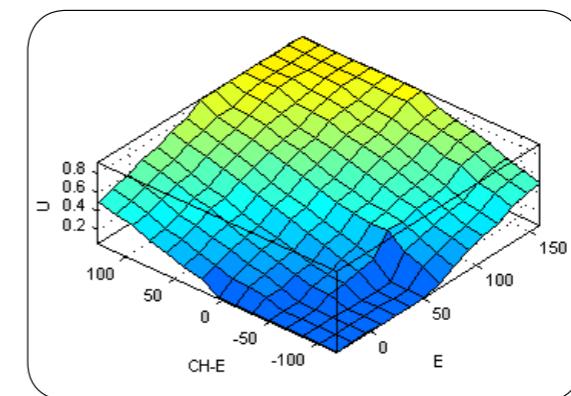


Fig. (6): 3D view of the FLC response

6. Self Tuning Pid Controller

Its need to automatically adjust PID gain parameters for obtaining the best response for short time because the Conventional PID controller not given desired performance for systems with uncertain dynamics and non-

linearity [9]. Fig. (6) shows general block simulation of Self-tuning fuzzy PID_controller .The automatic tuning for PID gain has been done using fuzzy logic techniques. Figure.7 shows the complete Simulink/MATLAB block for STF PID controller [12].

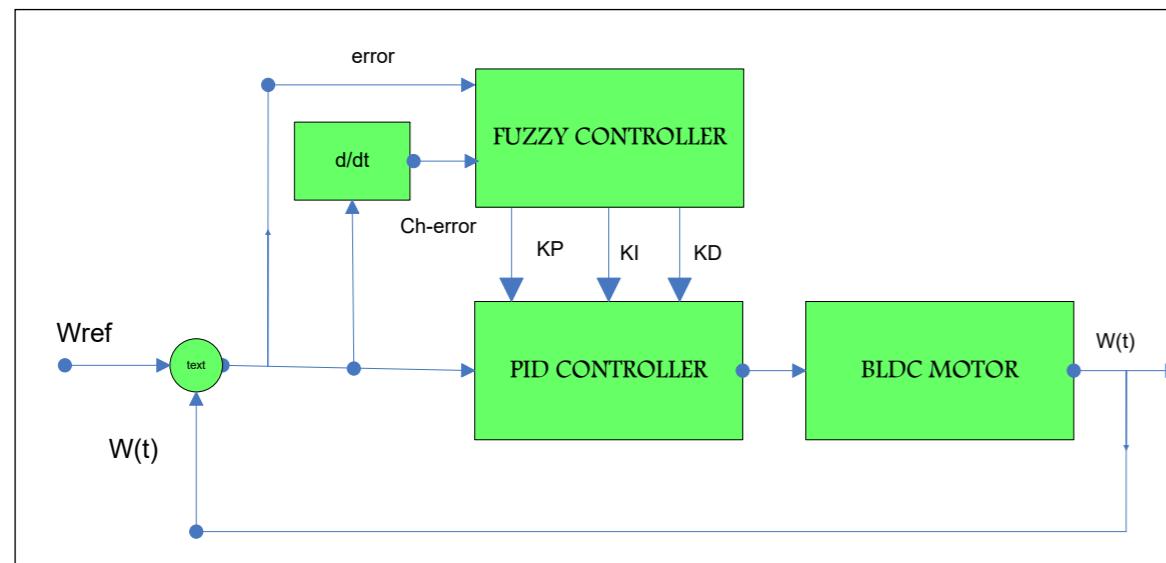


Fig. (7): Block Simulation of STF PID controller

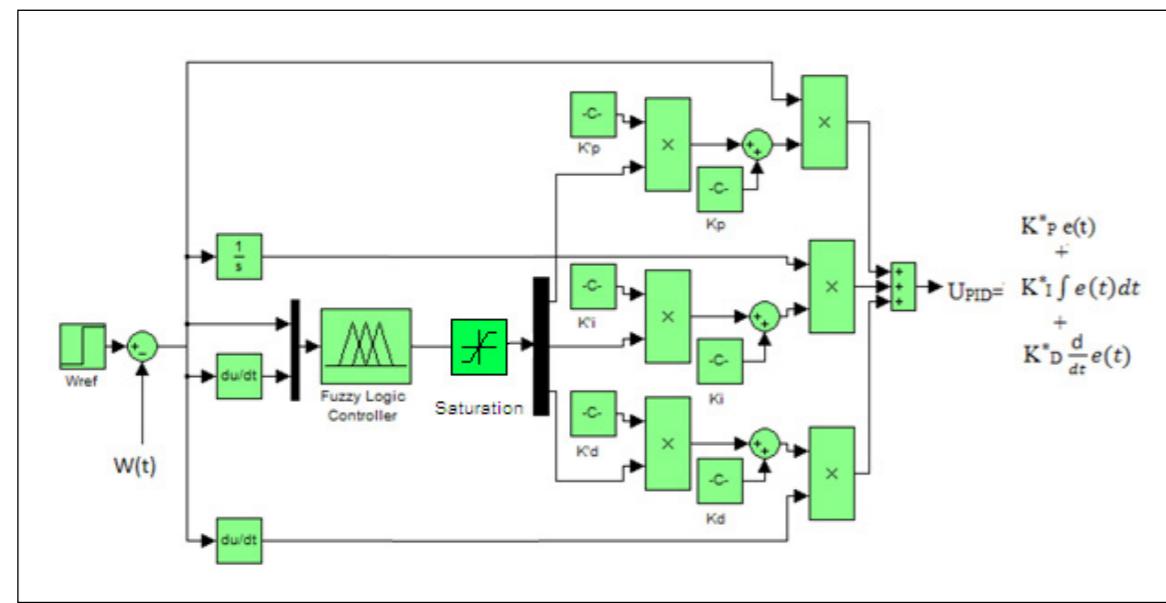


Fig. (8): Simulink/Model of STF PID controller

The STFPID transfer function that used in this work is :

$$U^{PID} = K_p^* \text{error}(t) + K_i^* \int \text{error}(t)dt + K_d^* \frac{d}{dt} \text{error}(t) \quad (3)$$

where :

(K_p^*, K_i^*, K_d^*) are the PID gain parameters that can be tuned. The adaptation equation for PID parameters can be given as [13]:

$$K_p^* = K_p' + K_p$$

$$K_i^* = K_i' + K_i$$

$$K_d^* = K_d' + K_d$$

Where (K_p', K_i', K_d') are the outputs of the fuzzy controller that are varying with time and we can be calculated it as:

$$(K_p') = (K_p - K_{p\min}) / (K_{p\max} - K_{p\min}) \quad (4)$$

$$(K_i') = (K_i - K_{i\min}) / (K_{i\max} - K_{i\min}) \quad (5)$$

$$(K_d') = (K_d - K_{d\min}) / (K_{d\max} - K_{d\min}) \quad (6)$$

Where: (K_p, K_i, K_d) are the initial values of the PID gain [13,14].

The variable parameters of the PID gains is $((K_{p\min}, K_{p\max}), (K_{i\min}, K_{i\max}) \& (K_{d\min}, K_{d\max}))$ respectively. We determined the parameter width of each PID gain by using the simulation result as shown in fig.(9). fig.(10) shows the response of (P,I and D) of error to determine the PID range gains [15].

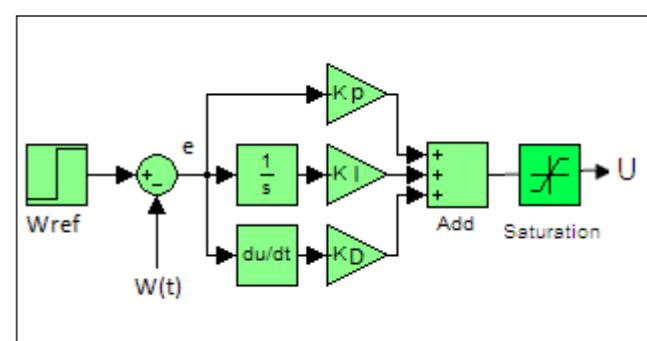


Fig. (9): Simulation part of PID controller parameters

It was taken the maximum and minimum variable ranges of the PID controller gains as shown below:

$$K_p \in [-10, 65], K_i \in [0, 1, 3] \text{ and } K_d \in [-820, 130].$$

Therefor can be calculated equations 4, 5 and 6 as follows[16]:

$$(K_p') = (K_p + 10) / 75$$

$$(K_i') = (K_i - 0, 1) / 2,9$$

$$(K_d') = (K_d + 820) / 950$$

Tuning membership functions and tuning their own elements, and the amount of overlap between them to obtain the suitable value for the PID gain [17].The inputs to controller are (e) and (ch-e) while the output is self-tuning fuzzy PID gain(K_p, K_i, K_d). Structure of fuzzy-PID controller with two input-three output structure implemented in Matlab as shown in Fig.(10) [18].

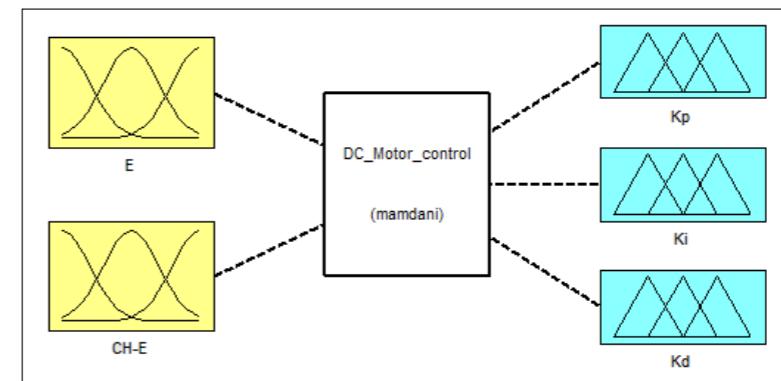


Fig.(10): Fuzzy inference block

The most important parts in design of fuzzy logic system are the inputs and outputs membership functions, and the if-then rules knowledge. The

three input- output tables corresponded to PID gains parameters, which are shown as table (3), table (4) and table (5).

Table (3): K'P Fuzzy rule table

CH-ERROR								
ERROR	<u>K'</u> _P	<u>L</u> _b	<u>L</u> _m	<u>L</u> _s	<u>Z</u>	<u>H</u> _s	<u>H</u> _m	<u>H</u> _b
	<u>L</u> _b	<u>H</u> _b	<u>H</u> _b	<u>H</u> _m	<u>H</u> _m	<u>H</u> _s	<u>Z</u>	<u>Z</u>
	<u>L</u> _m	<u>H</u> _b	<u>H</u> _b	<u>H</u> _m	<u>H</u> _s	<u>H</u> _s	<u>Z</u>	<u>Z</u>
	<u>L</u> _s	<u>H</u> _m	<u>H</u> _m	<u>H</u> _m	<u>H</u> _m	<u>Z</u>	<u>Z</u>	<u>L</u> _s
	<u>Z</u>	<u>H</u> _m	<u>H</u> _m	<u>H</u> _s	<u>Z</u>	<u>L</u> _s	<u>L</u> _s	<u>L</u> _m
	<u>H</u> _s	<u>H</u> _s	<u>H</u> _s	<u>Z</u>	<u>L</u> _s	<u>L</u> _s	<u>L</u> _m	<u>L</u> _m
	<u>H</u> _m	<u>H</u> _s	<u>Z</u>	<u>L</u> _s	<u>L</u> _m	<u>L</u> _m	<u>L</u> _m	<u>L</u> _b
	<u>H</u> _b	<u>Z</u>	<u>Z</u>	<u>L</u> _m	<u>L</u> _m	<u>L</u> _m	<u>L</u> _b	<u>L</u> _b

Table (4): K'I Fuzzy rule table

CH-ERROR								
ERROR	<u>K'</u> _I	<u>L</u> _b	<u>L</u> _m	<u>L</u> _s	<u>Z</u>	<u>H</u> _s	<u>H</u> _m	<u>H</u> _b
	<u>L</u> _b	<u>L</u> _b	<u>L</u> _b	<u>L</u> _m	<u>L</u> _m	<u>L</u> _s	<u>Z</u>	<u>Z</u>
	<u>L</u> _m	<u>L</u> _b	<u>L</u> _b	<u>L</u> _m	<u>L</u> _s	<u>L</u> _s	<u>Z</u>	<u>Z</u>
	<u>L</u> _s	<u>L</u> _b	<u>L</u> _m	<u>L</u> _s	<u>L</u> _s	<u>Z</u>	<u>H</u> _s	<u>H</u> _s
	<u>Z</u>	<u>L</u> _m	<u>L</u> _m	<u>L</u> _s	<u>Z</u>	<u>H</u> _s	<u>H</u> _m	<u>H</u> _m
	<u>H</u> _s	<u>L</u> _m	<u>L</u> _s	<u>Z</u>	<u>H</u> _s	<u>H</u> _s	<u>H</u> _m	<u>H</u> _b
	<u>H</u> _m	<u>Z</u>	<u>Z</u>	<u>H</u> _s	<u>H</u> _s	<u>H</u> _m	<u>H</u> _b	<u>H</u> _b
	<u>H</u> _b	<u>Z</u>	<u>Z</u>	<u>H</u> _s	<u>H</u> _m	<u>H</u> _m	<u>H</u> _b	<u>H</u> _b

Table (5): K'D Fuzzy rule table

CH-ERROR								
ERROR	<u>K'</u> _D	<u>L</u> _b	<u>L</u> _m	<u>L</u> _s	<u>Z</u>	<u>H</u> _s	<u>H</u> _m	<u>H</u> _b
	<u>L</u> _b	<u>H</u> _s	<u>L</u> _s	<u>L</u> _b	<u>L</u> _b	<u>L</u> _b	<u>L</u> _m	<u>H</u> _s
	<u>L</u> _m	<u>H</u> _s	<u>L</u> _s	<u>L</u> _s	<u>L</u> _b	<u>L</u> _b	<u>L</u> _b	<u>L</u> _m
	<u>L</u> _s	<u>Z</u>	<u>L</u> _s	<u>L</u> _s	<u>L</u> _m	<u>L</u> _m	<u>L</u> _m	<u>Z</u>
	<u>Z</u>	<u>Z</u>	<u>L</u> _s	<u>Z</u>				
	<u>H</u> _s	<u>Z</u>						
	<u>H</u> _m	<u>H</u> _b	<u>L</u> _s	<u>L</u> _s	<u>H</u> _s	<u>H</u> _s	<u>H</u> _s	<u>H</u> _b
	<u>H</u> _b	<u>H</u> _b	<u>H</u> _m	<u>H</u> _m	<u>H</u> _m	<u>H</u> _m	<u>H</u> _s	<u>H</u> _b

The STF-controller has two input (error & change of error) and three output variables (K'P, K'I& K'd) [19]. The Rule viewer got the parameters from Fuzzy PID controller is shown in figure(11), there are 49 rules are defined [20,21].

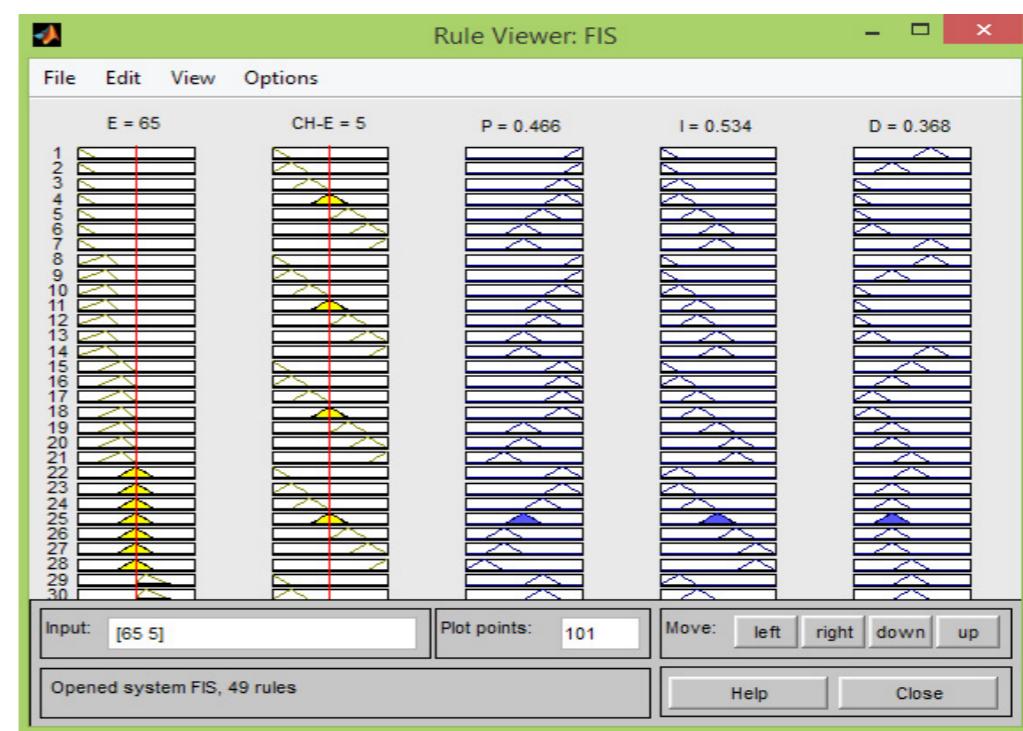


Fig. (11): The Rule viewer of the Self-tuning Fuzzy Controller

6. RESULT & DISCUSSION

In this work we presents a compared to different study of speed control system technique for the BLDC motor. To declaration the performance of three controllers, It is compared with design conventional PID controller , FLC and (STFPID) controller. Fig.

(12) shows the BLDC motor step speed responses with three different controllers technique in no load, where we was chosen the reference speed at 60(rpm) without applied external torque load. Fig. (12) show the speed response of the BLDC motor with three controllers with no load conditions,

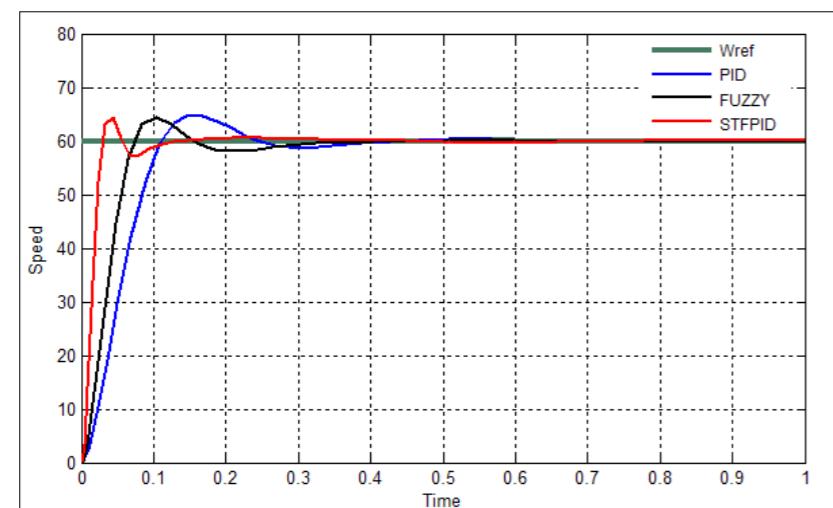


Fig. (12): The three different controllers responses

The parameters gain of the conventional PID controller that are used in this controller are ($P=1.764$, $I=19.74$ and $D=0.0274$) . From responses shown in Fig. (15), note that self- tuning fuzzy PID

control gives the better performance compared to other controllers. Table VII lists the performance of BLDC Motor with three controllers.

Table (6): Performance Comparison of PID , FL and STFPID Controller

Controller Type	Rise Time (ms)	Settling Time (ms)	Mo.sh (%)	Pick Time (ms)
PID	0.0745	0.325	8.12	1.08
Fuzzy	0.0487	0.263	7.32	1.07
STFPID	0.0191	0.156	7.41	0.54

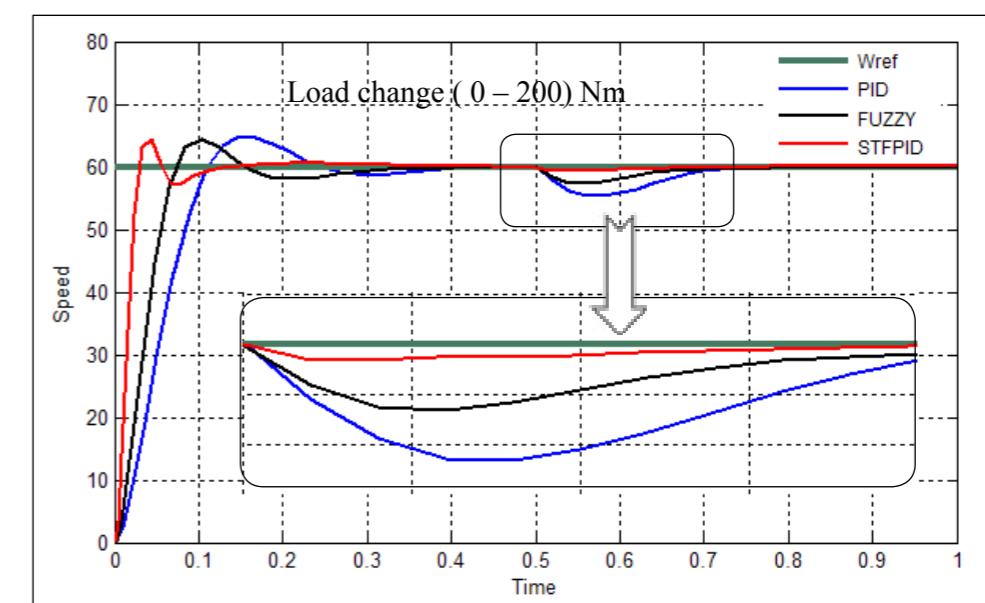


Fig. (13): The Speed Response for BLDC Motor with external load step change from 0 to 20 Nm at time = 0.5 sec.

From simulation results, It is found that the self-tuning fuzzy PID gain controller is the best in that it cancelation external torque disturbances comparative with other controller. The ability controllers to achieving the target speed of BLDC Motor speed is one of the important aspects of this

study. Fig. (14) shows the speed responses of the DC motor with three different controllers technique, when we changed the desired speed from (60 - 120) r.p.m . Fig. (19) shows the speed responses characteristics of the BLDC motor speed with pulses change in desired speed from - 60 to 60 rpm.

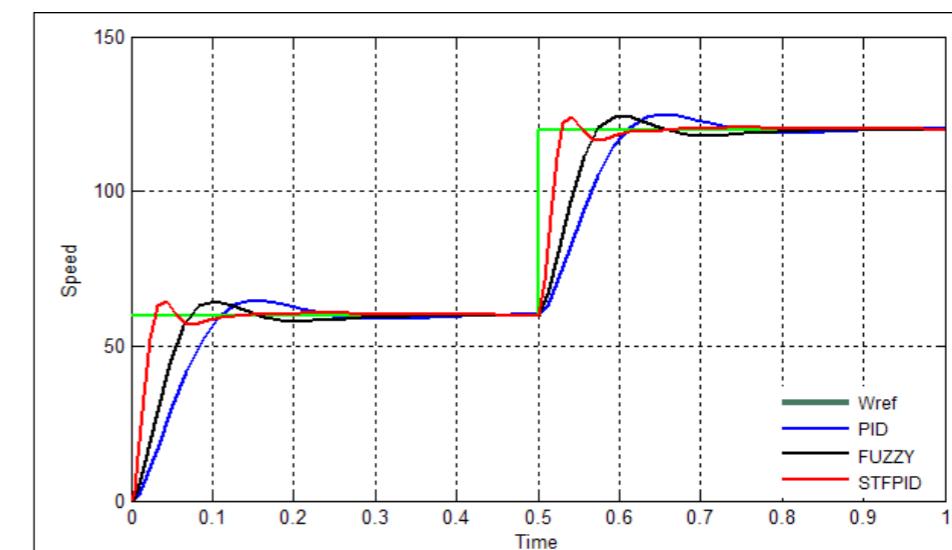


Fig. (14): Step change desired speed at t=0.5 sec from 60 - 120 rpm.

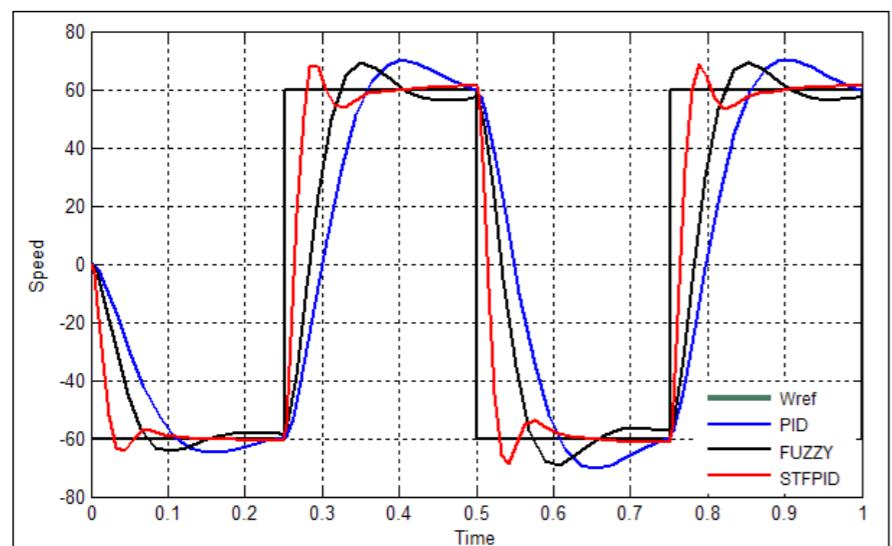


Fig. (15): DC motor speed with pulses change in reference speed from -60 to 60 rpm

From figs. (14 and 15), it is noted that the speed response with the STFPID controller is the faster to achieve the reference speed compared with other controllers.

8. Conclusion

In this work, the optimal self-tuning PID system is designed and simulated with automatically tuning PID controller parameters(KP,KI& KD).

According to the MATLAB/Simulink result shown in table VII, it is obvious at the same condition of operate the speed response of BLDC motor using (STFPID) technique given best performance than the pure PID-controller and Fuzzy logic system. In addition, the speed of the motor is constant during change the load and robust for cancelation the external load disturbances. Finally, it is found that the (STFPID) Controller gives a better results.

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Lower Bound Estimates for Approximation by Neural Networks for Functions in L_p Spaces for $p < 1$.

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الخلاصة

قدمنا في هذا البحث نظرية مباشرة للتقرير باستخدام الشبكات العصبية للدوال في الفضاءات L_p ، عندما $1 < p$.

الكلمات المفتاحية

الشبكات العصبية، مقياس النعومة، مبرهنة مباشرة.